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Time-Scale Synthesis of Non-Stationary Signals

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- 1 Introduction : nonstationarity
- 2 Locally deformed signals : an analysis-based approach
- 3 Locally deformed signals : a synthesis-based approach
- 4 Conclusion and perspectives

Stationarity

Definition (Stationarity)

A random process X is said to be second-order stationary if :

$$\blacksquare \mathbb{E}\{X(t)\} = m_X, \ \forall t ,$$

$$\blacksquare \mathbb{E}\{X(t)X(\tau)\} = k_X(t-\tau), \ \forall (t,\tau) \ .$$

Spectrum :

- Gives the distribution over frequencies of the power of X.
- Many methods to estimate the spectrum from a single realization of the stationary process X (e.g. Welch method).

Introduction			
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Nonstationarity			

The stationarity assumption is often irrelevant to study real-life signals, such as audio signals, or physiological signals.

$\Rightarrow \textbf{Questions}:$

- 1 Which classes of nonstationarity should we consider?
- 2 How should we analyze nonstationarity? In particular, how to extend spectral estimation to nonstationary signals?

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Broken stationarity : a class of nonstationarity¹

Two key ingredients :

- **1** A zero-mean stationary random process X.
- 2 A deformation operator that breaks stationarity \mathcal{T} .

We observe the "deformed" process Y given by :

 $Y = \mathcal{T}X$.

 \Rightarrow We limit ourselves to some physically relevant forms of operators.

^{1.} H. Omer. Modèles de déformation de processus stochastiques généralisés. Application à l'estimation des non stationnarités dans les signaux audio. PhD thesis, Aix-Marseille Université, 2015

Introduction		
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Deformation o	perators	

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Amplitude modulation

$$\mathcal{A}_{lpha}: \qquad \mathcal{A}_{lpha} x(t) = lpha(t) x(t) \; ,$$

with $\alpha \in C^1$ such that $\forall t, \ 0 < c_{\alpha} \leq \alpha(t) \leq C_{\alpha} < \infty$.



Frequency modulation

Introduction		
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Deformatio	n operators	

Time warping

$$\mathcal{D}_\gamma: \qquad \mathcal{D}_\gamma x(t) = \sqrt{\gamma'(t)} x(\gamma(t)) \;,$$

where $\gamma \in \mathcal{C}^2$ is a strictly increasing function such that

$$0 < c_\gamma \leq \gamma'(t) \leq C_\gamma < \infty, orall t$$
 .



Any combination of the above deformations

From a single realization of the nonstationary random process Y, we aim at estimating simultaneously :

- The spectrum \mathscr{S}_X of the underlying stationary random process X,
- The deformation operator \mathcal{T} .

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1 Introduction : nonstationarity

2 Locally deformed signals : an analysis-based approach

- Model
- Wavelet transform and approximation
- Estimation algorithm : JEFAS
- Applications to audio signals

3 Locally deformed signals : a synthesis-based approach

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Model and goal

Model :

 $Y=\mathscr{A}_{\alpha}\mathscr{D}_{\gamma}X\ .$

where X is a stationary process.

- Relevant to model physical phenomena, such as Doppler effect, speed variations, or animal vocalizations.
- **Goal** : From a single realization of the process *Y*, estimate simultaneously :
 - the spectrum \mathscr{S}_X of the underlying stationary process X,
 - the deformation functions α and γ .

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Wavelet transform

Definition (Wavelet transform)

$$\mathcal{W}_{\mathsf{x}}(s, au) = \langle \mathsf{x},\psi_{s au}
angle$$
 avec $\psi_{s au}(t) = 2^{-s/2}\psi\left(2^{-s}(t- au)
ight)$

where ψ is the analysis wavelet.



FIGURE – "Sharp wavelet" for two different values of s.

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Approximated behavior



Approximation theorem

The wavelet transforms of X and Y are approximately related by :

$$\mathcal{W}_{Y}(s, au)pprox \widetilde{\mathcal{W}}_{Y}(s, au)=lpha(au)\mathcal{W}_{X}\left(s+\log_{2}(\gamma'(au)),\gamma(au)
ight)\;.$$

The error term $\epsilon = W_Y - W_Y$ is a zero-mean random process, whose variance $\mathbb{E}\left\{ |\epsilon(s, \tau)|^2 \right\}$ depends on the regularity of α and γ' , and the speed of decay of ψ .

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Estimation	procedure	

Fix $\tau \Rightarrow$ Unknown parameters :

• \mathscr{S}_X • $\theta_1 = \alpha(\tau)^2$ • $\theta_2 = \log_2(\gamma'(\tau))$

Assumption : X is a zero-mean stationary Gaussian process. \Rightarrow Each column of the wavelet transform of $Y : \mathbf{w}_{Y,\tau} \sim C\mathcal{N}_c(0, \mathbf{C}(\Theta))$, with covariance matrix :

$$\mathbf{C}(\Theta)_{ij} = \theta_1 2^{(s_i+s_j+2\theta_2)/2} \int_0^\infty \mathscr{S}_{\mathbf{X}}(\xi) \overline{\hat{\psi}}(2^{s_i+\theta_2}\xi) \hat{\psi}(2^{s_j+\theta_2}\xi) d\xi \,.$$

 \Rightarrow The log-likelihood is given by

$$\mathscr{L}(\Theta) = -\frac{1}{2} \ln |\det(\mathbf{C}(\Theta))| - \frac{1}{2} \mathbf{C}(\Theta)^{-1} \mathbf{w}_{y,\tau} \cdot \mathbf{w}_{y,\tau}$$

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Estimation algorithm : JEFAS

The JEFAS (*Joint Estimation of Frequency, Amplitude and Spectrum*) algorithm consists in an alternating estimation.

Initializations :

- Initialize the power spectrum estimate.
- Initialize the amplitude modulation by a constant.
- $k \leftarrow 1$

while stopping criterion = FALSE do

- Time warping : Estimate $\tilde{\alpha}^{(k)}$ by ML, $\forall \tau$.
- Amplitude modulation : Estimate $\tilde{\gamma}^{(k)}$ by ML, $\forall \tau$.
- Spectrum : Estimate $\tilde{\mathscr{I}}_{\chi}^{(k)}$ from the "rectified" wavelet transform.
- $k \leftarrow k+1$

end while

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Doppler effect		

Assumptions :

- A source emits a stationary sound.
- The source follows a uniform linear motion, at speed V.
- From a fixed station, we record the sound emitted by the source.



 \Rightarrow Due to the Doppler effect, the sound we receive is time-warped, with :

$$\gamma'(t) = \frac{c^2}{c^2 - V^2} \left(1 - \frac{V^2 t}{\sqrt{d^2(c^2 - V^2) + (cVt)^2}} \right)$$



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Doppler effect



Comparison to the theoretical function with : d = 5 m and V = 54 m/s.

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Spectral analysis of a broadband wind sound



FIGURE – Top : Scalograms of the original signal (left) and the estimated stationary signal (right). Bottom left : estimated time warping and amplitude modulation. Bottom right : estimated spectrum.

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2 Locally deformed signals : an analysis-based approach

3 Locally deformed signals : a synthesis-based approach

- Motivations and model
- Estimation algorithm : JEFAS-S
- Illustrations

4 Conclusion and perspectives

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 Locally harmonic signal

Signal of the form :

 $y(t) = A_1 \cos(2\pi\xi_1\gamma(t)) + A_2 \cos(2\pi\xi_2\gamma(t)) \; ,$

where γ' is the fast varying instantaneous frequency.



■ JEFAS : Interference patterns on the wavelet transform. ⇒ Approximated behavior does not hold.

 \Rightarrow JEFAS diverges.

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Spectral estimation : Analysis vs. Synthesis

1 Analysis-based approach \Rightarrow JEFAS and JEFAS-BSS

Model of nonstationarity : locally time-warped signals, of the form :

 $Y = \mathscr{D}_{\gamma} X \; ,$

where X is an arbitrary stationary process.

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Spectral estimation : Analysis vs. Synthesis

1 Analysis-based approach \Rightarrow JEFAS and JEFAS-BSS

Model of nonstationarity : locally time-warped signals, of the form : $= \mathscr{D}_{\gamma} X ,$ where X is an arbitrary stationary process. Introduction JEFAS JEFAS-Synthesis Conclusion and perspectives

Spectral estimation : Analysis vs. Synthesis

2 Synthesis-based approach \Rightarrow JEFAS-S

Synthesis model \Leftrightarrow Reconstruction formula :

$$y(t) = \operatorname{Re}\left(\sum_{s} (\psi_{s} * W_{s})(t)\right) + \epsilon(t) ,$$

where $W_s(t)$ are random time-scale coefficients, and $\epsilon(t)$ is a noise.

► Discretization of the problem :

$$\mathbf{y} = \mathsf{Re}\left(\sum_{n=1}^{N} \mathbf{\Psi}_{n} \mathbf{w}_{n}\right) + \boldsymbol{\epsilon} \; ,$$

where \mathbf{w}_n is the *n*-th column of the time-scale representation.

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Bavesian infe	rence		

• Likelihood : $\boldsymbol{\epsilon} \sim \mathcal{N}(\mathbf{0}, \, \sigma^2 \mathbf{I})$

$$p(\mathbf{y}|\mathbf{w}_n) = \mathcal{N}\left(\operatorname{Re}\left(\sum_{n=1}^{N} \Psi_n \mathbf{w}_n\right), \sigma^2 \mathbf{I}\right)$$

Prior : on the synthesis coefficients w_n

Uncorrelated vectors such that :

$$\mathbf{w}_n \sim \mathcal{CN}_c(\mathbf{0}, \mathbf{C}_n)$$
.

Covariance matrices C_n are translated versions of reference covariance function c :

$$[\mathbf{C}_n]_{mm'} = [\mathbf{C}(\theta_n)]_{mm'} = \mathbf{c}(\mathbf{s}_m + \theta_n, \mathbf{s}_{m'} + \theta_n) ,$$

where $\theta_n \in \mathbb{R}$ is the shift parameter.

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Ectimation et	ratamu		

Estimation strategy

- **Expectation Maximization (EM) principle where :**
 - \bullet is the parameter,
 - \mathbf{w}_n is the latent variable.

EM steps

The update at iteration k relies on the following two steps :

■ Time-scale representation update

Maximum a posteriori estimation :

$$\tilde{\mathbf{w}}_{n}^{(k)} = rac{1}{2} \mathbf{C} \left(\widetilde{ heta}_{n}^{(k-1)}
ight) \Psi_{n}^{H} \mathbf{C}_{y} \left(\widetilde{ heta}^{(k-1)}
ight)^{-1} \mathbf{y} \; .$$

2 Nonstationarity parameter update :

$$\tilde{\theta}_n^{(k)} = \arg \max_{\theta_n} \ \mathscr{L}(\theta_n) - \frac{1}{2} \mathrm{Tr} \left(\mathsf{C}(\theta_n)^{-1} \mathbf{\Gamma}_n \left(\tilde{\boldsymbol{\theta}}^{(k-1)} \right) \right) \ ,$$

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Algorithm : JEFAS-S

Initialization : estimate $\tilde{\theta}^{(0)}$ and $\tilde{\mathscr{S}}^{(0)}$ using JEFAS.

- $k \leftarrow 1$
- while stopping criterion = FALSE do
 - Time-scale representation estimation : $\tilde{\mathbf{w}}_{n}^{(k)}$.
 - Time-warping parameter estimation : $\tilde{ heta}^{(k)}$.
 - Spectrum estimation : $\tilde{\mathscr{S}}_{X}^{(k)}$.
 - $k \leftarrow k+1$.

end while

- Alternating estimation \Rightarrow Similar to JEFAS.
- Convergence ensured by the EM principle.
- Additional estimation of the time-scale representation \Rightarrow JEFAS-S is slower to converge than JEFAS.

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Estimated adapted

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Broadband synthetic signal



- Time-warping parameter estimation : not improved with respect to JEFAS.
- Allows denoising : improvement of **7.06 dB** of the Signal to Noise Ratio between the measurements \mathbf{y} and the reconstructed signal $\tilde{\mathbf{y}}_0$.

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Locally ha	rmonic signal		

Signal of the form :

$y(t) = A_1 \cos(2\pi \xi_1 \gamma(t)) + A_2 \cos(2\pi \xi_2 \gamma(t)) ,$

where γ' is the (normalized) fast varying instantaneous frequency.

JEFAS-S : Prior of uncorrelation between $\mathbf{w}_n \Rightarrow No$ interference \Rightarrow JEFAS-S converges.





Time (s)

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Time (s)

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Conclusion

Summary :

Broken stationarities :

$$Y = \mathcal{T}X$$

- Locally deformed signals
- Multivariate locally deformed signals
- Locally harmonic signals
- Spectral estimation : Simultaneous estimation of the spectrum \mathscr{S}_X and the deformation operator \mathcal{T} .

Two estimation strategies :

- Analysis-based approaches
- Synthesis-based approaches



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Formula 1

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